## Electromagnetic Energy Conversion ELEC0431

## Exercise session 1: Phasors and power in the sinusoidal steady state

9 February 2024

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## In this class...

Today's aim: Introduce a very important concept: The phasors (pay particular attention to this class!)
$>$ Reminders (one-ports and Kirchhoff's laws)
$>$ Harmonic functions
$>$ Root Mean Square value
$>$ The phasors
$>$ Exercise 1
$>$ Product of two harmonic functions
$>$ Active, reactive and apparent power
$>$ Exercises 2

## Reminders

One-ports
Kirchhoff laws

## One-ports (or "dipoles" in French)

A one-port is a two-terminal electric component.
It can be:

Terminal 1
$>$ A resistance $R$
$>$ An inductance $L$
$\Rightarrow$ A capacitance $C$
$>$ A power generator
$>$ A diode
$>$ A combination of components
$>$ Etc.


Terminal 2

## One-ports (or "dipoles" in French)

Each one-port is associated with a current $i(t)$ and a voltage $v(t)$.
$>i(t)$ is positive if the positive charges flow in the direction indicated by the arrows.
$>v(t)$ is positive if the tension at the head of the arrow is larger than the one at the tip.
Four different configurations can thus be obtained.
$\rightarrow$ Two of them consume power and two of them deliver power.

Consume
Power
(like a passive component)


Deliver
Power
(like an active component)
$\rightarrow$ This leads to two different conventions: the passive convention and the active convention.

## Passive VS Active conventions

## The passive convention

(also called receiver, motor or load convention)
The passive one-port receives the power $p(t)=v(t) i(t)$


This convention is used for $R, L, C$, lumped parameters, motors and primary side of transformers.

## The active convention

(also called generator or source convention)
The active one-port delivers the power $p(t)=v(t) i(t)$


This convention is used for voltage sources, current sources, generator and secondary side of transformers.

## $v(t)-\mathrm{i}(\mathrm{t})$ relationship of basic components



## Kirchhoff's laws

## Kirchhoff's first law

## (also called Kirchhoff's junction rule or Kirchhoff's current law)

At any junction in the electrical circuit, the sum of currents flowing into the junction is equal to the sum of currents flowing out of the junction.

$\bigcirc \quad 0=i_{A}(t)+i_{B}(t)+i_{D}(t)$
$\bigcirc i_{B}(t)=i_{C}(t)+i_{E}(t)$
$\bigcirc i_{A}(t)+i_{C}(t)=i_{F}(t)$

- $i_{D}(t)+i_{E}(t)+i_{F}(t)=0$


## Kirchhoff's laws

## Kirchhoff's second law

## (also called Kirchhoff's loop rule or Kirchhoff's voltage law)

Around any closed loop in a circuit, the directed sum of potential differences across components is zero.


$$
\begin{aligned}
& \longrightarrow v_{D}(t)-v_{B}(t)-v_{E}(t)=0 \\
& \longrightarrow v_{E}(t)-v_{C}(t)-v_{F}(t)=0
\end{aligned}
$$

$$
\longrightarrow v_{B}(t)-v_{A}(t)+v_{C}(t)=0
$$

## Introduction to phasors

## Harmonic functions

Root Mean square value
The phasors
Harmonic function $\Leftrightarrow$ Phasor equivalence
Phasors applied to circuits
Application: Exercise 1

## Harmonic functions

In this class, we deal with harmonic functions:

$$
f(t)=F_{p e a k} \cos (\omega t+\theta)
$$

where:


The angular frequency is linked to the frequency $f$ and period $T$ as

$$
\omega=2 \pi f=\frac{2 \pi}{T}
$$

## Root Mean square (RMS) value

Let's consider a resistance $R$ :


The instantaneous power $p(t)$ dissipated in the resistance is

$$
p(t)=v(t) i(t)=R i(t)^{2}=\frac{v(t)^{2}}{R}
$$

And the mean dissipated power is

$$
P=\frac{1}{T} \int_{0}^{T} v(t) i(t) d t=R \underbrace{\frac{1}{T} \int_{0}^{T} i(t)^{2} d t}_{I_{R M S}^{2}}=\frac{1}{R} \underbrace{\frac{1}{T} \int_{0}^{T} v(t)^{2} d t}_{V_{R M S}^{2}}
$$

## Root Mean square (RMS) value

The Root Mean Square (RMS) value (or effective value) of $f(t)$ is defined as:

$$
F_{R M S}=\sqrt{\frac{1}{T} \int_{0}^{T} f(t)^{2} d t} .
$$

With $f(t)=F_{\text {peak }} \cos (\omega t+\theta)$, we obtain:

$$
\begin{aligned}
F_{R M S} & =\sqrt{\frac{1}{T} \int_{0}^{T} F_{\text {peak }}^{2} \cos ^{2}(\omega t+\theta) d t}=F_{\text {peak }} \sqrt{\frac{1}{T} \int_{0}^{T} \cos ^{2}(\omega t) d t} \\
& =F_{p e a k} \sqrt{\frac{1}{T} \int_{0}^{T} \frac{1+\cos (2 \omega t)}{2} d t}=F_{\text {peak }} \sqrt{\frac{1}{T} \int_{0}^{T} \frac{1}{2} d t}=\frac{F_{\text {peak }}}{\sqrt{2}}
\end{aligned}
$$

## Root Mean square (RMS) value

## Remark:

The following conventions will be used in the rest of the course for RMS and peak/max values:

$$
\begin{array}{rll}
F_{R M S} & \Rightarrow & F \\
F_{\text {peak }} & \Rightarrow & F_{m}
\end{array}
$$

In this class, we deal only with harmonic functions $f(t)=F_{m} \cos (\omega t+\theta)$ for which we have

$$
F_{m}=\sqrt{2} F
$$

## The phasors

The definition of the phasors is obtained by manipulating the expression of $f(t)$ :

$$
\begin{aligned}
f(t) & =F_{m} \cos (\omega t+\theta) \\
& =\sqrt{2} F \cos (\omega t+\theta) \\
& =\boldsymbol{R}\left(\sqrt{2} F e^{j(\omega t+\theta)}\right) \\
& =\boldsymbol{R}(\underbrace{\sqrt{2} F e^{j \theta}}_{\bar{F}} e^{j \omega t})
\end{aligned}
$$

$$
\bar{F}=F e^{j \theta}
$$

is the phasor of $f(t)$. It is a complex number with information on the signal amplitude $(F)$ and the signal phase $(\theta)$. For a system operating at a fixed frequency, $F$ and $\theta$ holds all the relevant information to describe the system.

## The phasors

For signals evolving at a same angular frequency $\boldsymbol{\omega}$, phasors are like a snapshot representation containing all the necessary information (phase angle and amplitude).

$$
\begin{array}{lll}
f_{1}(t)=\sqrt{2} F_{1} \cos \left(\omega t+\theta_{1}\right) & \Leftrightarrow & \bar{F}_{1}=F_{1} e^{j \theta_{1}}=F_{1} \angle \theta_{1} \\
f_{2}(t)=\sqrt{2} F_{2} \cos \left(\omega t+\theta_{2}\right) & \Leftrightarrow & \bar{F}_{2}=F_{2} e^{j \theta_{2}}=F_{2} \angle \theta_{2}
\end{array}
$$



## Harmonic function $\Leftrightarrow$ Phasor equivalence

Linearity:

$$
\left.\begin{array}{rl}
f_{1}(t)+f_{2}(t) & =F_{1 m} \cos (\omega t+\theta)+F_{2 m} \cos (\omega t+\psi) \\
& =\sqrt{2} F_{1} \cos (\omega t+\theta)+\sqrt{2} F_{2} \cos (\omega t+\psi) \\
& =\boldsymbol{R}\left(\sqrt{2} F_{1} \mathrm{e}^{j(\omega t+\theta)}\right)+\boldsymbol{R}\left(\sqrt{2} F_{2} e^{j(\omega t+\psi)}\right) \\
& =\boldsymbol{R}(\sqrt{2} e^{j \omega t}(\underbrace{F_{1} e^{j \theta}+F_{2} e^{j \psi}}_{\bar{F}_{1}+\bar{F}_{2}})
\end{array}\right)
$$

## Harmonic function $\Leftrightarrow$ Phasor equivalence

Differentiation:

$$
\begin{aligned}
\frac{d f(t)}{d t} & =\frac{d}{d t}\left(F_{m} \cos (\omega t+\theta)\right) \\
& =\frac{d}{d t} \boldsymbol{R}\left(\sqrt{2} F e^{j(\omega t+\theta)}\right) \\
& =\boldsymbol{R}\left(\sqrt{2} F e^{j(\omega t+\theta)} j \omega\right) \\
& =\boldsymbol{R}(\sqrt{2} \underbrace{j \omega F e^{j \theta}}_{j \omega \bar{F}} e^{j \omega t})
\end{aligned}
$$

$$
\frac{d f(t)}{d t} \quad \Leftrightarrow \quad j \omega \bar{F}
$$

## Harmonic function $\Leftrightarrow$ Phasor equivalence

## Integration:

$$
\begin{aligned}
\int f(t) d t & =\int F_{m} \cos (\omega t+\theta) d t \\
& =\int \boldsymbol{R}\left(\sqrt{2} F e^{j(\omega t+\theta)}\right) d t \\
& =\boldsymbol{R}\left(\sqrt{2} F e^{j(\omega t+\theta) 1 / j \omega)}\right. \\
& =\boldsymbol{R}(\sqrt{2} \underbrace{1 / j \omega F e^{j \theta}}_{\bar{F} / j \omega} e^{j \omega t})
\end{aligned}
$$

$$
\int f(t) d t \Leftrightarrow \frac{\bar{F}}{j \omega}
$$

## Phasors applied to circuits

The phasor formalism can be exported to voltages and currents:

$$
\begin{array}{rll}
v(t)=\sqrt{2} V \cos (\omega t+\theta) & \Leftrightarrow & \bar{V}=V e^{j \theta} \\
i(t)=\sqrt{2} I \cos (\omega t+\psi) & \Leftrightarrow & \bar{I}=I e^{j \psi}
\end{array}
$$




## Phasors applied to circuits

The phase angle difference (or phase lag) is noted $\varphi$. It is defined as:

$$
\varphi=\theta-\psi
$$

```
(\varphi always goes from the current to the voltage !!!)
```

It is common to take $\bar{V}$ as the reference with $\theta=0$ and $\varphi=-\psi$.



## Phasors applied to circuits

Time domain

$$
\begin{gathered}
f(t)=\sqrt{2} F \cos (\omega t+\theta) \\
a f_{1}(t)+b f_{2}(t) \\
d f(t) / d t \\
\int f(t) d t
\end{gathered}
$$

$$
v(t)=R i(t)
$$

$$
v(t)=L d i(t) / d t
$$

$$
v(t)=(1 / C) \int i(t) d t
$$

$$
\begin{array}{lc}
\Leftrightarrow & \text { Frequency don } \\
\Leftrightarrow & \bar{F}=F e^{j \theta} \\
\Leftrightarrow & a \bar{F}_{1}+b \bar{F}_{2} \\
\Leftrightarrow & j \omega \bar{F} \\
\Leftrightarrow & \bar{F} / j \omega \\
\Leftrightarrow & \bar{V}=R \bar{I} \\
\Leftrightarrow & \bar{V}=j \omega L \bar{I} \\
\Leftrightarrow & \bar{V}=\bar{I} / j \omega C
\end{array}
$$

Thanks to the linearity equivalence, the Kirchhoff laws can be applied to phasors. In addition, resistances, inductances and capacitances can be replaced by complex impedances $Z=\frac{\bar{V}}{\bar{I}}$ of value respectively equal to $R, j \omega L$ and $\frac{1}{j \omega C}$.

## Application: Exercise 1

For the hereunder circuit powered with a 50 Hz sinusoidal voltage generator, find the voltages and currents across all components (magnitude and phase angle). Represent all of them on a phasor diagram.


## Application: Exercise 1: Watch out !



The sum of the norms of individual complex numbers usually differs from the norm of the sum of those complex numbers.

## Homework 1

For the hereunder circuit powered with a 50 Hz sinusoidal voltage generator, find the voltages and currents across all components (magnitude and phase angle). Represent all of them on a phasor diagram.


## Homework 2

For the hereunder circuit powered with a 200 Hz sinusoidal voltage generator, find the voltages and currents across all components (magnitude and phase angle). Represent all of them on a phasor diagram.


## Homework 3

For the hereunder circuit powered with a 50 Hz sinusoidal voltage generator, find the voltages and currents across all components (magnitude and phase angle). Represent all of them on a phasor diagram.


## Homework 4

For the hereunder circuit powered with a 200 Hz sinusoidal voltage generator, find the voltages and currents across all components (magnitude and phase angle). Represent all of them on a phasor diagram.


## Homework 5

For the hereunder circuit powered with a 50 Hz sinusoidal voltage generator, find the voltages and currents across all components (magnitude and phase angle). Represent all of them on a phasor diagram.


## Homework 6

A resistor $\mathrm{R}=10 \Omega$ and an inductor L are connected in series to a 50 Hz sinusoidal voltage generator of RMS voltage $V=230 \mathrm{~V}$. An RMS voltage of 100 V is measured on the terminals of the resistor. What is the value of L?

Hint:
Use the norm of $R+j \omega L$.


## Homework 7

Consider the circuit hereunder operating with a pulsation $\omega=2 \pi f=100 \mathrm{rad} / \mathrm{s}$. Find the voltages and currents across all components (magnitude and phase angle). Represent all of them on a phasor diagram. Do the same in case the inductance is increased to 200 mH .



## Homework 8

Consider the circuit hereunder operating at 50 Hz . What is the value of $C$ if the RMS current is measured to be $20 A$ ? Find the voltages and currents across all components (magnitude and phase angle). Represent all of them on a phasor diagram.


## Homework 9

Consider the load showed on the right.


To determine the value of the resistance and the value of the inductance, two tests are performed:

- A DC voltage of 150 V is applied between the load terminals for a measured current of 1.95 A .
- An AC voltage of RMS value 230 V , oscillating at 50 Hz , is applied between the load terminals for a measured RMS current of 2.81 A .

What are the values of $R$ and $L$ ?

## Homework 10

Considering the circuit depicted hereunder, find the input voltage $\bar{V}$ (magnitude and phase angle).


Active, reactive, complex and apparent power

Product of two harmonic functions Active, reactive, complex and apparent power Application: Exercise 2

## Product of two harmonic functions

The instantaneous power $p(t)$ consumed by passive components (or delivered by active components) is the product of the instantaneous voltage $v(t)$ with the instantaneous current $i(t)$ :

$$
\begin{aligned}
p(t) & =v(t) i(t) \\
& =\sqrt{2} V \cos (\omega t+\theta) \\
& =2 V I \cos (\omega t+\theta) \\
& =V I[\underbrace{\cos (\theta-\psi)}_{\text {DC }}+\underbrace{\cos (2 \omega t+\theta+\psi)}_{\begin{array}{c}
\text { AC component } \\
\cos (\omega t) \\
\cos (\omega t+\psi) \\
\text { cosponent }
\end{array}} \begin{array}{c}
\text { frequency) } \\
\text { frequal }
\end{array}
\end{aligned}
$$

## Product of two harmonic functions



$$
p(t)=V I[\cos (\theta-\psi)+\cos (2 \omega t+\theta+\psi)]
$$

The instantaneous power can be split in two parts:
$>\mathrm{VI} \cos (\theta-\psi): \mathrm{A} \mathrm{DC} \mathrm{component}$ called the Active power $P$.
$>V I \cos (2 \omega t+\theta+\psi): \mathrm{An} \mathrm{AC}$ component oscillating at twice the original frequency.

## Active, reactive, complex and apparent power

$$
\begin{aligned}
p(t) & =v(t) i(t) \\
& =V I[\cos (\theta-\psi)+\cos (2 \omega t+\theta+\psi)] \\
& =V I[\cos (\theta-\psi)+\cos (2(\omega t+\theta)-(\theta-\psi))] \\
& =V I[\cos (\theta-\psi)+\cos (2(\omega t+\theta)) \cos (\theta-\psi)+\sin (2(\omega t+\theta)) \sin (\theta-\psi)] \\
& =\underbrace{V I \cos (\theta-\psi)}_{P} \underbrace{1+\cos (2(\omega t+\theta))]}_{P}+\underbrace{V I \sin (\theta-\psi)}_{\geq 0} \underbrace{\sin (2(\omega t+\theta))}_{Q}
\end{aligned}
$$

$p(t)$ can be decomposed into:
$>$ A flow of energy going always in the same direction with an average value $P=V I \cos (\theta-\psi)$. The value $P$ is called the active power.
$>$ A fluctuating flow of energy, exchanged back and forth twice a period, with an average value $Q=V I \sin (\theta-\psi)$. The value $Q$ is called the reactive power.

## Active, reactive, complex and apparent power

Although $p(t)=P[1+\cos (2(\omega t+\theta))]+Q \sin (2(\omega t+\theta))$ is a real number, one defines the complex power $S$ as:

$$
S=P+j Q
$$

$S$ IS NOT A PHASOR, it is just a convenient complex number allowing to describe easily and completely the instantaneous power (real number).

One also shows that is is really easy to determine $S$ using phasors:

$$
\begin{aligned}
S & =P+j Q=V I \cos (\theta-\psi)+j V I \sin (\theta-\psi)=V I[\cos (\theta-\psi)+j \sin (\theta-\psi)] \\
& =V I e^{j(\theta-\psi)}=V e^{j \theta} I e^{-j \psi}=\bar{V} \bar{I}^{*}
\end{aligned}
$$



## Active, reactive, complex and apparent power

One also defines the apparent power:

$$
\begin{aligned}
|S| & =\sqrt{P^{2}+Q^{2}} \\
& =\sqrt{V^{2} I^{2} \cos ^{2}(\theta-\psi)+V^{2} I^{2} \sin ^{2}(\theta-\psi)} \\
& =V I
\end{aligned}
$$

And the power factor:

$$
P F=\cos (\theta-\psi)=\cos (\varphi)=\frac{P}{|S|}
$$

Let's now suppose that we want to deliver an active power $P$ to a load. What current $I$ should we inject for a fixed voltage $V$ ?

$$
I=\frac{|S|}{V}=\frac{P}{V \cos (\varphi)}
$$

$I$ is larger when $\cos (\varphi)$ is small $\rightarrow$ If $\cos (\varphi)$ is small, there are more transmission losses.
$\rightarrow$ We want $\cos (\varphi)$ to be $1 \rightarrow$ To avoid transmission losses, we want the reactive power $Q$ to be as small as possible.

## Application: Exercise 2

Your colleague suggests to add a $50 \mu F$ capacitor in parallel of the RL load. It is supposed to compensate the reactive power consumed by the inductive load. Is it a good idea ? If so, what would be the exact value of $C$ needed ?


## Homework 11

Fill the cells of the table below with the most appropriate answer among

$$
=0,<0, \quad>0, \quad=1, \quad<1, \quad+\infty \text { and }-\infty .
$$

|  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| one-port: |  |  |  |  |  |
| active power consumed |  |  |  |  |  |
| reactive power produced |  |  |  |  |  |
| $\cos \phi$ |  |  |  |  |  |
| $\tan \phi$ |  |  |  |  |  |

