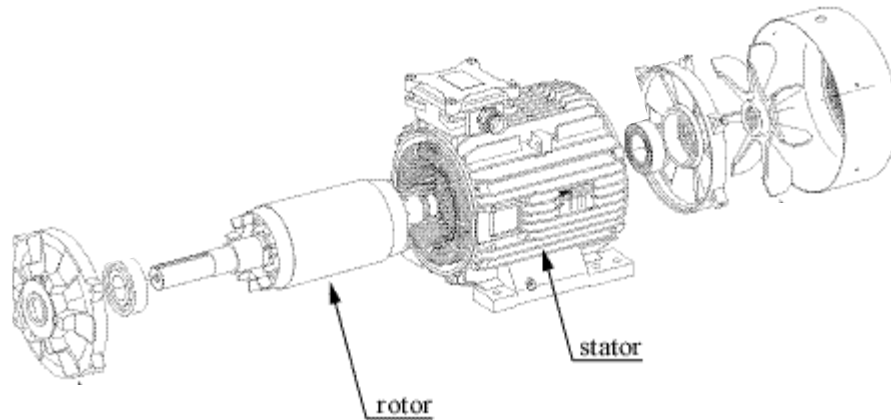


Asynchronous machines

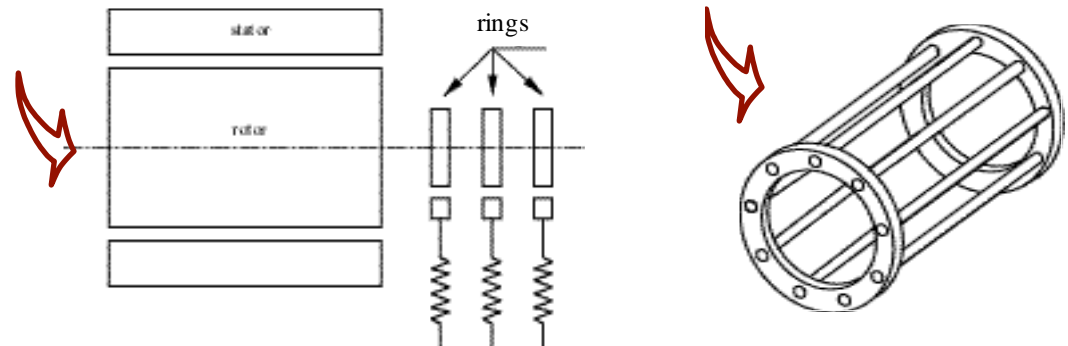


Stator: magnetic circuit and polyphased winding (usually 3-phase), with p pairs of poles, with polyphased currents

Rotor: magnetic circuit and electric circuit...

*... or made of copper or aluminum bars, short-circuited at each extremity of the rotor → **squirrel-cage rotor***

*... either made of a polyphased winding, (with star connection in order to avoid currents between rotor phases) connected to short-circuit rings on the rotor axis; brushes connect these rings to external resistances → **wound rotor***



General principle

Wound rotor machine initially at rest ...

$$g = 1$$

$$\begin{array}{l} \text{def} \\ g = \text{slip} \end{array}$$

When 3-phase currents of pulsation ω flow in the stator windings they create a rotating magnetic flux density with p pairs of poles, rotating with angular speed ω/p .

Rotation speed

$$\dot{\theta} = (1 - g) \frac{\omega}{p}$$

$$0 \leq g \leq 1$$

This field induces 3-phased e.m.f.s of pulsation ω in the (stationary) rotor windings, and thus 3-phase currents of pulsation ω .

Angular speed of the stator rotating field with respect to the rotor

$$\frac{\omega}{p} - \dot{\theta} = \frac{g \omega}{p}$$

e.m.f.s and induced currents with pulsation $g\omega$, and torque

The interaction between the rotating stator flux density and the induced currents in the rotor creates a torque, which puts the rotor into movement

$$g = 0$$

Synchronous speed

$$\dot{\theta}_s = \frac{\omega}{p}$$

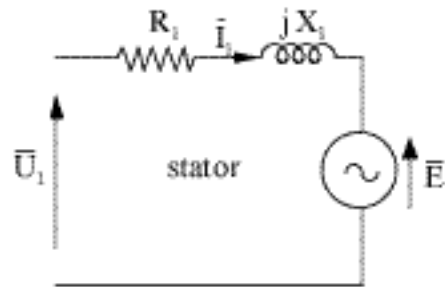
(asynchronous machine, or induction machine)

no e.m.f., no current, no torque

General equations

Stator equation

Rotating magnetic flux density in the airgap leads to e.m.f. \bar{E}_1 in stator winding, with pulsation ω

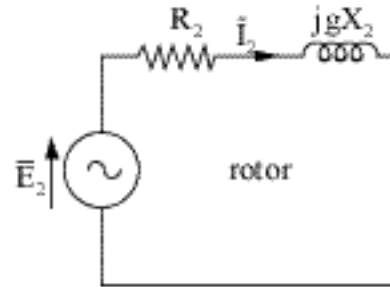


(a)

For one phase

Rotor equation

Rotating magnetic flux density in the airgap leads to e.m.f. \bar{E}_2 in rotor winding, with pulsation ω .



(b)

$$\bar{U}_1 = \bar{E}_1 + (R_1 + jX_1) \bar{I}_1$$

$X_1 =$ Leakage reactance of one stator phase
 $R_1 =$ Resistance of a stator phase

$$\bar{E}_2 = (R_2 + jgX_2) \bar{I}_2$$

$X_2 =$ Leakage reactance of a rotor phase at pulsation ω
 $R_2 =$ Resistance of a rotor phase

$$\bar{E}_1 = -j\omega k_1 \bar{\Phi}_r$$

$$\bar{E}_2 = -jg\omega k_2 \bar{\Phi}_r$$

Link equation

resulting m.m.f.

$$\bar{F}_r = \bar{F}_1 + \bar{F}_2 = -k'_1 \bar{I}_1 + k'_2 \bar{I}_2$$

with

$$\bar{\Phi}_r = \frac{\bar{F}_r}{R_{ep}}$$

resulting flux

$$\frac{\bar{E}_1}{\bar{E}_2} = \frac{n_{tr}}{g}$$

$$k_1/k_2 = k'_1/k'_2 = n_{tr} \dots$$

$$\bar{E}_1 = jX_\mu \left(\bar{I}_1 - \frac{\bar{I}_2}{n_{tr}} \right)$$

$X_\mu =$ Magnetizing reactance

Equivalent circuit

$$\bar{U}_1 = \bar{E}_1 + (R_1 + jX_1) \bar{I}_1 \quad \text{(a)}$$

$$\frac{\bar{E}_2}{g} = \left(\frac{R_2}{g} + jX_2 \right) \bar{I}_2$$

$$\frac{\bar{E}_1}{\bar{E}_2} = \frac{n_{tr}}{g}$$

$$\bar{E}_1 = jX_\mu \left(\bar{I}_1 - \frac{\bar{I}_2}{n_{tr}} \right)$$

$$\bar{E}_1 = \left(n_{tr}^2 \frac{R_2}{g} + jn_{tr}^2 X_2 \right) \frac{\bar{I}_2}{n_{tr}}$$

$$\bar{E}_1 = \left(\frac{R'_2}{g} + jX'_2 \right) \bar{I}'_2 \quad \text{(b)}$$

$$\bar{E}_1 = jX_\mu (\bar{I}_1 - \bar{I}'_2) \quad \text{(c)}$$

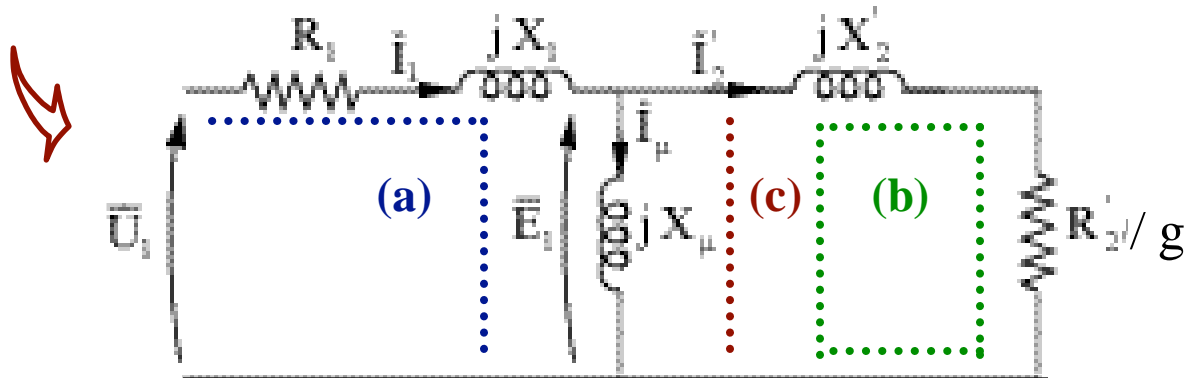
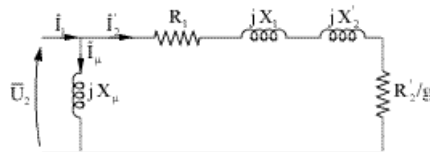
magnetizing current

$$\bar{I}_\mu = \bar{I}_1 - \bar{I}'_2$$

Simplified equivalent circuit

$$R_1, R'_2, X_1, X'_2 \ll X_\mu$$

(ratios ~ 40 X, 600 R)



Equivalent circuit parameters

Experimental determination of the parameters

No load test

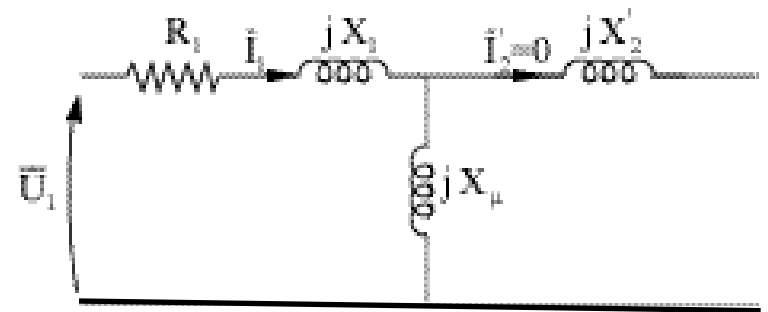
$$g = 0$$

Under nominal voltage and at synchronous speed

$$I_1 \ll \Rightarrow P_{\text{Joule stator}} \ll$$

$$P_v = P_{\text{mag}} \quad Q_v = 3 \frac{U_1^2}{X_\mu}$$

magnetic losses in the stator laminations



Short-circuit test

$$g = 1$$

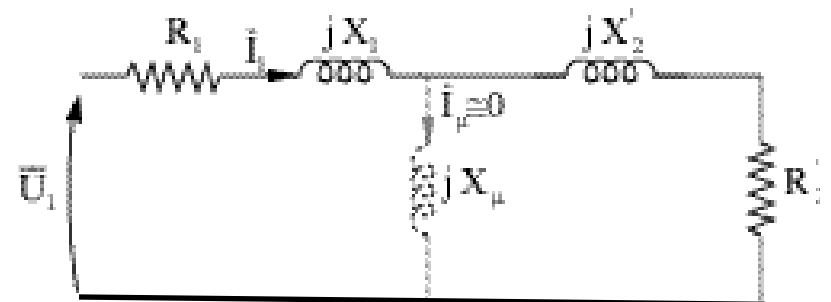
Under reduced voltage at zero speed (stationary rotor)

$$U_1 \ll \Rightarrow I_\mu \text{ and } P_{\text{mag}} \ll$$

$$P_{\text{cc}} = 3(R_1 + R'_2)I_1^2 \quad Q_{\text{cc}} = 3(X_1 + X'_2)I_1^2$$

*in the resistances
of the stator and rotor
windings*

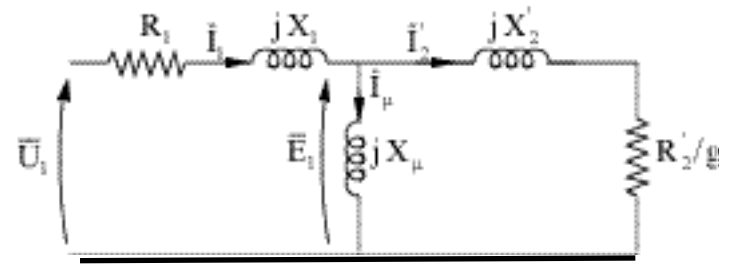
*in the leakage reactances
of the stator and the rotor*



Power, torque and efficiency

Electric power

$$P = 3 U_1 I_1 \cos \varphi = 3 \left(R_1 I_1^2 + \frac{R'_2}{g} I_2^2 \right)$$



Stator

Airgap

Rotor

Absorbed power

Power transmitted from stator to rotor

Electromagnetic power

Stator Joule losses $3 R_1 I_1^2$

$$P_{st \rightarrow rot} = P - 3 R_1 I_1^2 = 3 \frac{R'_2}{g} I_2^2$$

$$\eta_{stat} = \frac{P_{st \rightarrow rot}}{P}$$

Stator efficiency (1)

Rotor efficiency (2)

$$\eta_{rot} = \frac{P_{elm}}{P_{st \rightarrow rot}} = \frac{3 \frac{1-g}{g} R'_2 I_2^2}{3 \frac{1}{g} R'_2 I_2^2} = 1 - g$$

$$\eta_{mec} = \frac{P_{mec}}{P_{elm}}$$

Mechanical efficiency (3)

$$P_{elm} = P_{st \rightarrow rot} - 3 R'_2 I_2^2 = 3 \frac{1-g}{g} R'_2 I_2^2$$

Asynchronous motor efficiency

(1)x(2)x(3)

$$\eta = (1 - g) \eta_{stat} \eta_{mec} < 1 - g$$

$$C_{elm} = \frac{P_{elm}}{\dot{\theta}} = \frac{3 P R'_2}{\omega g} I_2^2$$

Electromagnetic torque

Asynchronous machines

Mechanical characteristic

Mechanical characteristic

Evolution of the electromagnetic torque C as a function of the slip g (or the rotation speed), for a given stator voltage U_1 and pulsation ω

$$C = \frac{3p}{\omega} \frac{R'_2}{g} I_2^2$$

$$C = f(g) \quad \text{with} \quad \begin{cases} U_1 = \text{constant} \\ \omega = \text{constant} \end{cases}$$

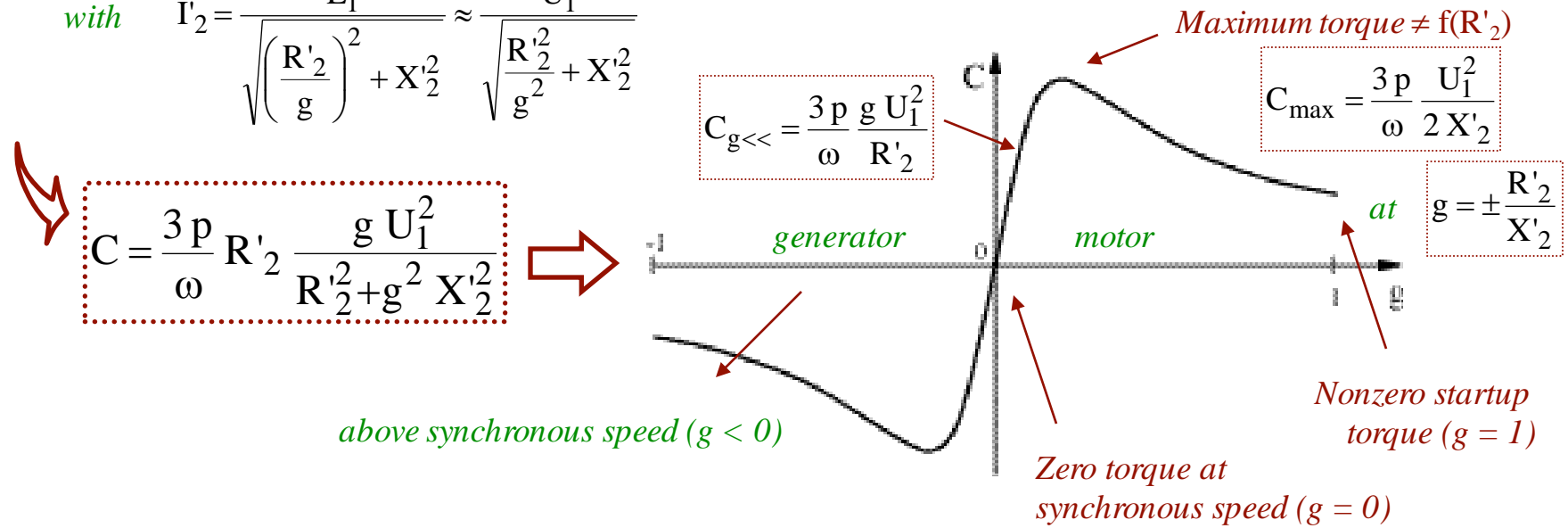
with
$$I_2 = \frac{E_1}{\sqrt{\left(\frac{R'_2}{g}\right)^2 + X_2'^2}} \approx \frac{U_1}{\sqrt{\frac{R_2'^2}{g^2} + X_2'^2}}$$

$$C = \frac{3p}{\omega} R'_2 \frac{g U_1^2}{R_2'^2 + g^2 X_2'^2}$$

$$C_{g \ll 1} = \frac{3p}{\omega} \frac{g U_1^2}{R'_2}$$

$$C_{\max} = \frac{3p}{\omega} \frac{U_1^2}{2 X_2'}$$

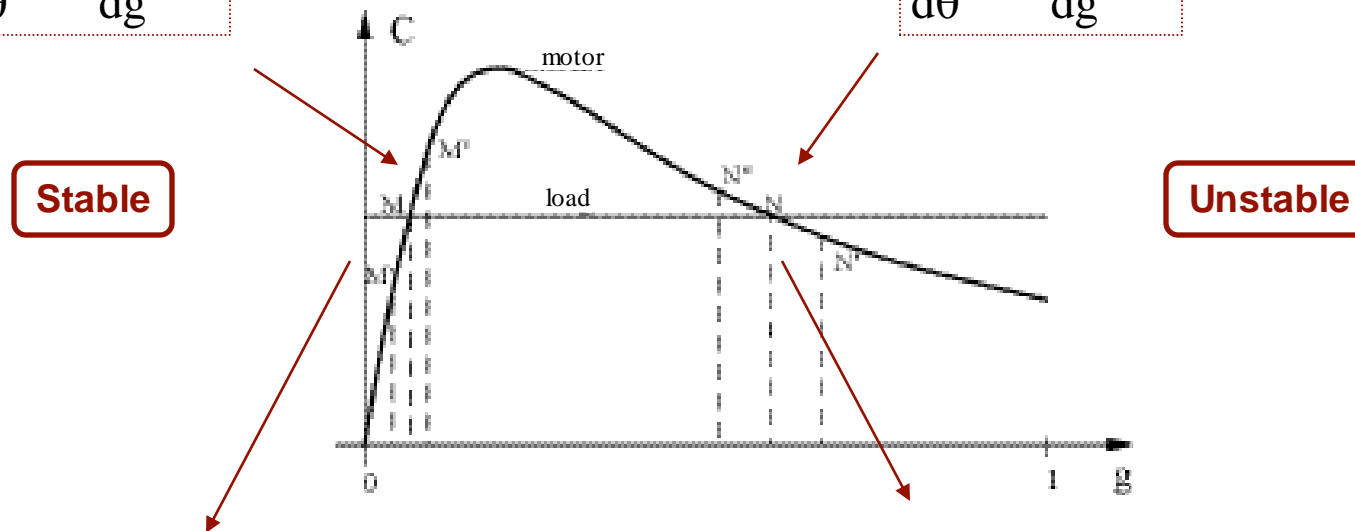
$$g = \pm \frac{R'_2}{X_2'}$$



Stability zone

$$\frac{dC}{d\dot{\theta}} = -\frac{dC}{dg} < 0$$

$$\frac{dC}{d\dot{\theta}} = -\frac{dC}{dg} > 0$$

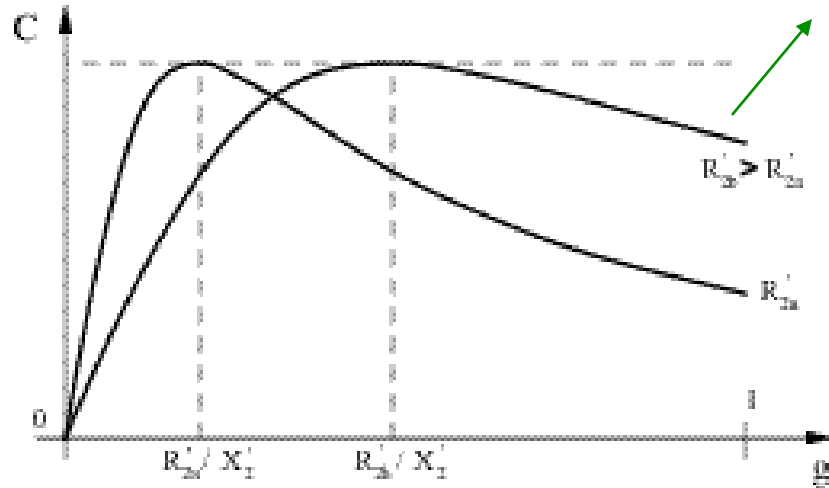


A decrease (resp. increase) in speed, or an increase (resp. decrease) in the slip g , leads to an increase (resp. decrease) of the motor torque, from point $M \rightarrow M'$ (resp. $M \rightarrow M''$). The machine will thus accelerate (resp. decelerate) to reach back point M , where the motor and resistant torques match
 \Rightarrow **stable**.

A decrease (resp. increase) in speed, or an increase (resp. decrease) in the slip g , leads to a decrease (resp. increase) of the motor torque, from point $N \rightarrow N'$ (resp. $N \rightarrow N''$). The machine will thus decelerate (resp. accelerate), further increasing the mismatch between the resistant and motor torques (resp. leading to an evolution towards the stable state $N'' \rightarrow M$)
 \Rightarrow **unstable**.

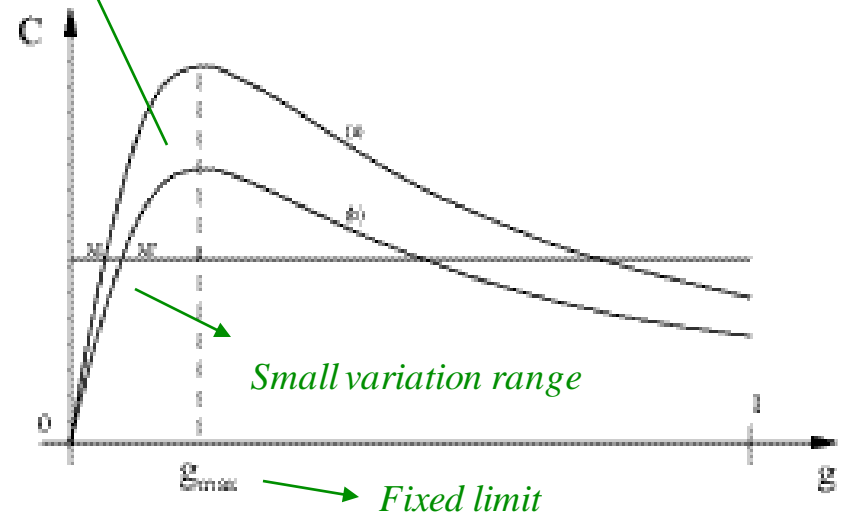
Speed control

Modification of the rotor resistance R'_2



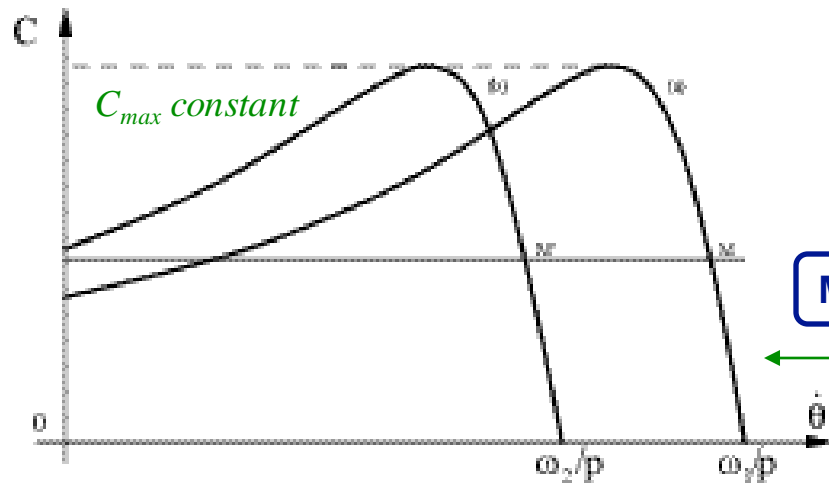
Reduction of speed at the cost of efficiency ($\eta < 1-g$)

Modification of the voltage U_1



Small variation range

Fixed limit



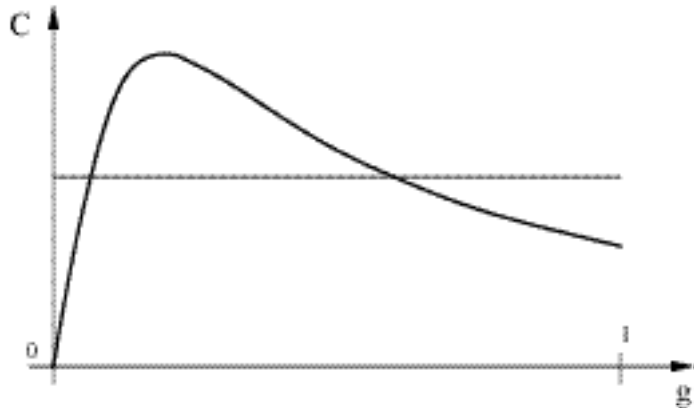
C_{max} constant

Modification of the frequency

Interest in keeping the magnetic flux (hence U_1/ω) constant...

$$\Phi_r = \lambda_\mu I_\mu = \frac{X_\mu I_\mu}{\omega} = \frac{E_1}{\omega} \approx \frac{U_1}{\omega}$$

Asynchronous motor startup



Startup torque (g = 1)

$$C = \frac{3p}{\omega} R'_2 \frac{g U_1^2}{R'^2_2 + X'^2_2}$$

usually quite small, since $R'_2 \ll$

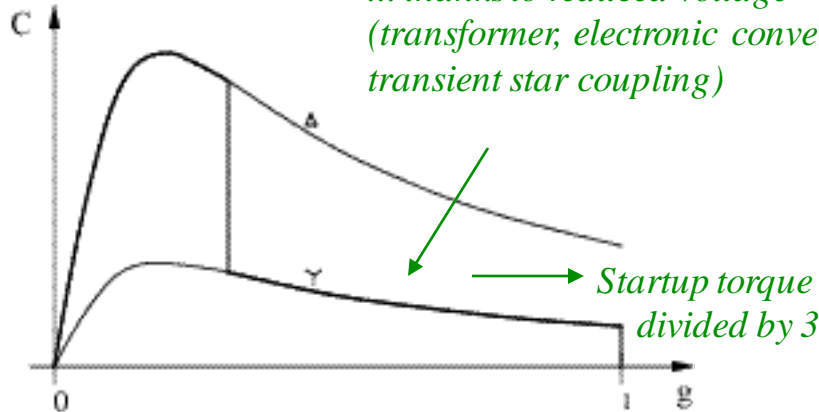
Startup current

$$I_d = \frac{U_1}{\sqrt{(R_1 + R'_2)^2 + (X_1 + X'_2)^2}}$$

very large, should be limited...

Squirrel cage motor

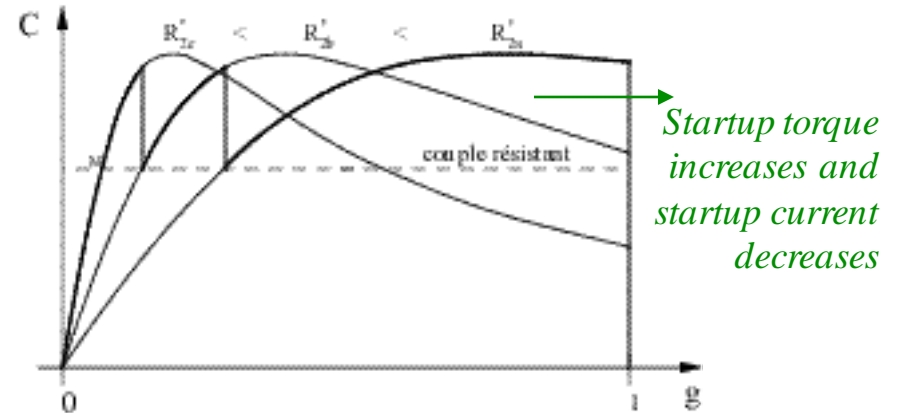
... thanks to reduced voltage (transformer, electronic converter, transient star coupling)



Startup possible only for small resistant torque at low speed (e.g. pumps, ventilators, air conditioners)

Wound rotor

... resistances in series with the rotor winding (progressively removed)



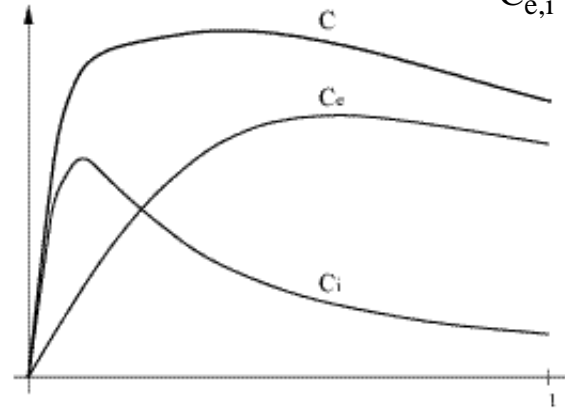
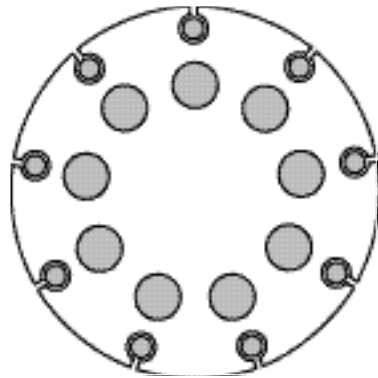
Special asynchronous motors

Double cage rotor

Exterior (e) and interior (i)

$$R'_e \gg R'_i$$

$$X'_e \ll X'_i$$



$$C_{e,i} = \frac{3p}{\omega} R'_{e,i} \frac{g U_1^2}{R_{e,i}^2 + g^2 X_{e,i}^2}$$

Startup ($g=1$):

$$R'_{i,e} \ll gX'_e \ll gX'_i \Rightarrow C_e \gg C_i$$

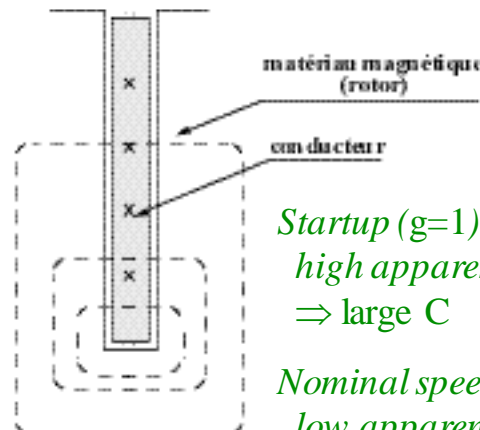
Nominal speed ($g \ll 1$):

$$R'_e \gg R'_i \gg gX'_{i,e} \Rightarrow C_i \gg C_e$$

Skin effect rotor

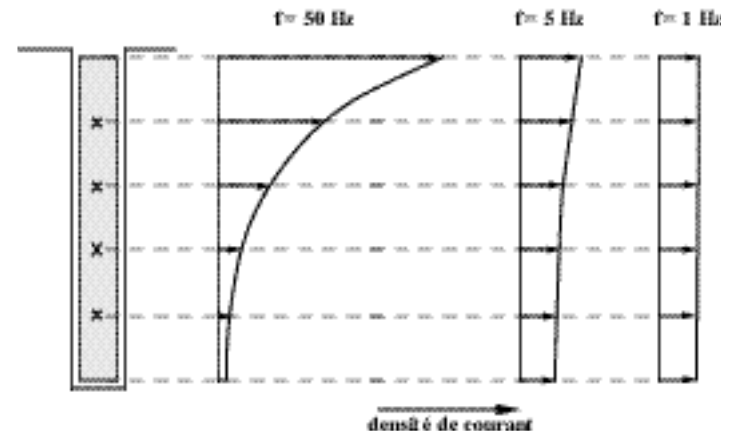
Elongated rectangular rotor conductors

\Rightarrow same behavior as double cage

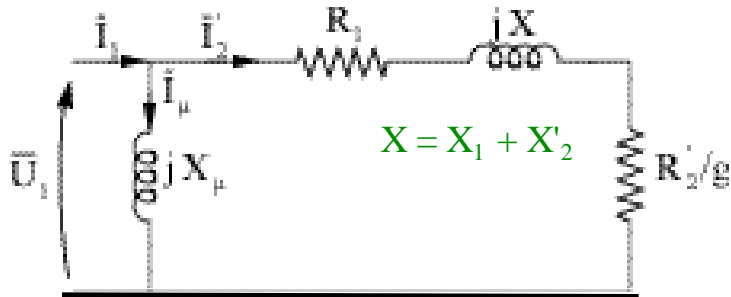


Startup ($g=1$):
high apparent resistance
 \Rightarrow large C

Nominal speed ($g \ll 1$):
low apparent resistance



Circle diagram



$$\bar{U}_1 = jX \bar{I}'_2 + (R_1 + R'_2/g) \bar{I}'_2$$

$$\bar{U}_1 = \text{constant} \quad (0)$$

$$\frac{\bar{U}_1}{jX} = \bar{I}'_2 + \frac{R_1 + R'_2/g}{jX} \bar{I}'_2$$

\Rightarrow Locus of \bar{I}'_2 is a circle of diameter U_1/X

Active power

$$P = 3 U_1 I_1 \cos \varphi = 3 U_1 AM$$

$$AM = \frac{P}{3 U_1} \quad (5)$$

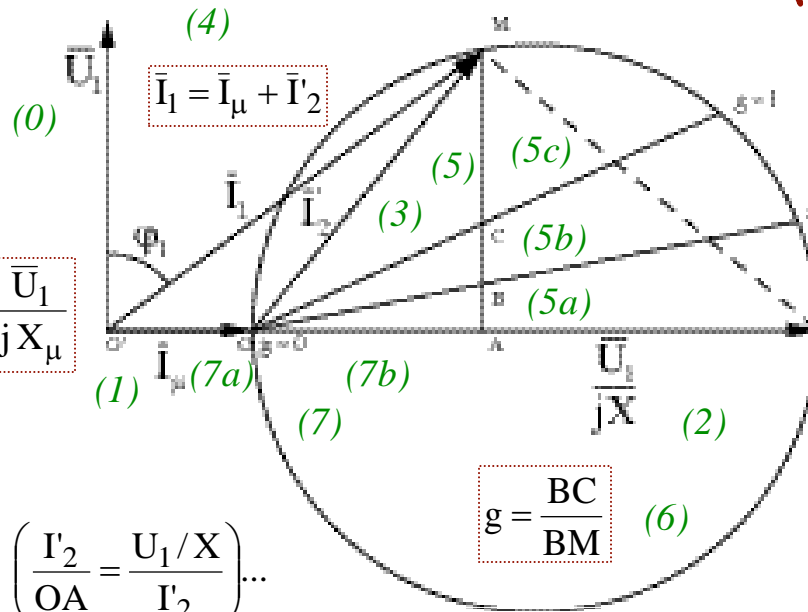
Reactive power

$$Q = 3 U_1 I_1 \sin \varphi = 3 U_1 O'A$$

$$O'A = \frac{Q}{3 U_1} \quad (7) = (7a) + (7b)$$

$$O'A = \frac{Q_{X_\mu} + Q_X}{3 U_1} = O'O + OA \left(\frac{I'_2}{OA} = \frac{U_1/X}{I'_2} \right) \dots$$

Reactive power in X_μ and X



Stator Joule losses

$$P_{Js} = 3 R_1 I_2'^2 = 3 R_1 OA \frac{U_1}{X}$$

$$AB = \frac{P_{Js}}{3 U_1} = \frac{R_1}{X} OA \quad (5a)$$

Rotor Joule losses

$$P_{Jr} = 3 R'_2 I_2'^2 = 3 R'_2 OA \frac{U_1}{X}$$

$$BC = \frac{P_{Jr}}{3 U_1} = \frac{R'_2}{X} OA \quad (5b)$$

Electromagnetic power

$$CM \quad (5c)$$

Operating modes

The machine

- absorbs electric power ($A_1M_1 \geq 0$) ;
- produces mechanical power ($C_1M_1 \geq 0$).

Asynchronous motor

The machine

- absorbs electric power ($A_3M_3 \geq 0$) ;
- absorbs mechanical power ($C_3M_3 \leq 0$).

Radiator

The machine

- produces electric power ($A_2M_2 \leq 0$) ;
- absorbs mechanical power ($C_2M_2 \leq 0$).

Asynchronous generator

